### 6.3 Signal Space Concepts

As in the case of vectors, we now develop a parallel treatment for a set of signals.

## Definition 6.34.

(a) The inner product of two real-valued signals $x_{1}(t)$ and $x_{2}(t)$ is denoted by $\left\langle x_{1}(t), x_{2}(t)\right\rangle$ and defined by

$$
\left\langle x_{1}(t), x_{2}(t)\right\rangle=\int_{-\infty}^{\infty} x_{1}(t) x_{2}(t) d t .
$$

(b) The signals are orthogonal if their inner product is zero.
(c) The norm of a signal is defined as

$$
\|x(t)\|=\sqrt{\langle x(t), x(t)\rangle}=\sqrt{E_{x}}
$$

where $E_{x}$ is the energy in $x(t)$ :

$$
\langle x(t), x(t)\rangle=\int_{-\infty}^{\infty}\left(e(t)(x(t)) d t=\int_{-\infty}^{\infty}|x(t)|^{2} d t \equiv E_{x}\right.
$$

(d) A collection of $N$ signals is orthonormal if the signals are orthogonal and their norms are all unity.

Example 6.35. Consider the two waveforms shown in Figure 26.




2
$=\int^{2} d t=\left.t\right|_{0} ^{2}=2$
Figure 26: Two Waveforms in Example 6.35
$\left\|s_{2}\right\|=\sqrt{2}$
$\left\|s_{1}\right\|=\sqrt{2}$

$$
\begin{aligned}
E_{A_{2}} & =E_{2} \\
& =\int_{-\infty}^{\infty}\left(A_{2}(t)\right)^{2} d t \\
& =2
\end{aligned}
$$

0

## Definition 6.36.

(a) The (orthogonal) projection of $x_{2}(t)$ to $x_{1}(t)$ is given by

$$
\operatorname{proj}_{x_{1}(t)} x_{2}(t)=\frac{\left\langle\boldsymbol{x}_{2}, x_{1}\right\rangle}{\left\langle\boldsymbol{x}_{1}, \boldsymbol{x}_{1}\right\rangle} \boldsymbol{x}_{1}=\frac{\left\langle x_{2}(t), x_{1}(t)\right\rangle}{\left\langle x_{1}(t), x_{1}(t)\right\rangle} x_{1}(t)=\frac{\left\langle x_{2}(t), x_{1}(t)\right\rangle}{E_{x_{1}}} x_{1}(t)
$$

(b) The cross-correlation coefficient of $x_{1}(t)$ and $x_{2}(t)$ is defined as

$$
\rho_{x_{1}, x_{2}}=\frac{\left\langle x_{1}(t), x_{2}(t)\right\rangle}{\sqrt{E_{x_{1}} E_{x_{2}}}} .
$$

- $\operatorname{proj}_{x_{1}(t)} x_{2}(t)=\sqrt{E_{x_{2}}} \rho_{x_{2}, x_{1}} \frac{x_{1}(t)}{\sqrt{E_{x_{1}}}}$

Example 6.37. For the two waveforms shown in Figure 26,

$$
\operatorname{proj}_{s_{,}(t)} s_{2}(t)=\frac{\left\langle s_{2}, s_{1}\right\rangle}{\left\langle s_{1}, s_{1}\right\rangle} s_{1}=\frac{0}{2} s_{1}=0
$$

6.38. Similar to 6.30, the Gram-Schmidt Orthogonalization Procedure (GSOP) can be used to construct a set of orthonormal waveforms from a set of finite energy signal waveforms: $\left\{s_{j}(t), j=1,2, \ldots, M\right\}$.

The first orthonormal function is simply constructed as

$$
\phi_{1}(t)=\frac{u_{1}(t)}{\sqrt{E_{u_{1}}}}=\frac{s_{1}(t)}{\sqrt{E_{s_{1}}}}
$$

The subsequent orthonormal functions are found as follows:

$$
\phi_{i}(t)=\frac{u_{i}(t)}{\sqrt{E_{u_{i}}}}
$$

where the unnormalized basis function $u_{i}(t)$ is given by

$$
u_{i}(t)=s_{i}(t)-\sum_{k=1}^{i-1} \operatorname{proj}_{u_{k}(t)} s_{i}(t)
$$

and

$$
\operatorname{proj}_{u_{k}(t)} s_{i}(t)=\frac{\left\langle s_{i}(t), u_{k}(t)\right\rangle}{\left\langle u_{k}(t), u_{k}(t)\right\rangle} u_{k}(t)=\left\langle s_{i}(t), \phi_{k}(t)\right\rangle \phi_{k}(t)
$$

As with the GSOP for vectors, we also discard the zero functions. In general, the final number of orthonormal functions, $N$, is less than or equal to the number of given waveforms, $M$, depending on one of the two possibilities:
(a) If the waveforms $\left\{s_{j}(t), j=1,2, \ldots, M\right\}$ form a linearly independent set, then $N=M$.
(b) If the waveforms $\left\{s_{j}(t), j=1,2, \ldots, M\right\}$ are not linearly independent, then $N<M$.

Example 6.39. Consider the four waveforms illustrated in Figure 27. Use the Gram-Schmidt orthogonalization procedure (where the waveforms are applied in the order given) to find the orthonormal basis waveforms $\phi_{1}(t)$, $\phi_{2}(t), \ldots$ whose linear combinations can be used to represent the four waveforms.



$$
\left\langle S_{3}, A_{1}\right\rangle=\int_{0}^{2} 1 \times 1 d t=2
$$

$$
\begin{aligned}
1 \uparrow & \left\langle s_{3}, s_{2}\right\rangle
\end{aligned} \begin{gathered}
1 \times 1 d t+\int_{0}^{2}(-1) \times 1 d t \\
\hline 0
\end{gathered}
$$

Figure 27: Four signals for orthogonalization in Example 6.39

$$
\text { Ex.6.35 } \quad \partial_{1}=\mu_{1}=\sqrt{E_{\mu},} \varnothing_{1} \quad=0
$$

- First we ssi; $u_{1}(t)=s_{1}(t)$. $\quad=\sqrt{2} \phi_{1}=\sqrt{2} \phi_{1}+0 \phi_{2}+o \phi_{3}$

$$
\begin{aligned}
& E_{u_{1}}=E_{s}\left(\frac{v_{2}}{2}\right. \\
& \phi_{1}(t)=-\frac{s_{1}(t)}{\sqrt{E_{u_{1}}}}=\frac{s_{1}(t)}{\sqrt{E_{s_{1}}}}=\frac{1}{\sqrt{2}} s_{1}(t) \\
& u_{2}(t)=s_{1}(t)-\operatorname{proj}_{u_{1}} s_{2}=s_{2}(t) \\
& (t) \\
& E_{2} .6 .37
\end{aligned}
$$

$$
\text { - } \begin{aligned}
& u_{2}(t)=s_{2}(t)-\underbrace{\operatorname{proj}_{u_{1}} s_{2}}=s_{2}(t) E_{x_{0} .6 .37} \\
& E_{u_{2}}=E_{s_{2}}=2
\end{aligned}
$$

$$
\phi_{2}(t)=\frac{u_{2}(t)}{\sqrt{E_{u_{2}}}}=\frac{1}{\sqrt{2}} u_{2}=\frac{1}{\sqrt{2}} \curvearrowright_{2}(t)
$$

$$
\partial_{2}=u_{2}=\sqrt{E_{u_{2}}} \varnothing_{2}=\sqrt{2} \phi_{2}
$$



$$
=O \Phi_{1}+\sqrt{2} \varnothing_{1}+O \varnothing_{3}
$$

$$
\begin{aligned}
& \text { - } u_{4}(t)=s_{4}(t)-\operatorname{proj}_{u_{1}} s_{4}-\underbrace{\operatorname{proj}_{u_{2}} s_{4}}-\underbrace{\operatorname{proj}_{u_{3}} s_{4}}
\end{aligned}
$$

$$
\begin{aligned}
& s_{4}=\sigma_{4}(t)-s_{1}(t)+\mu_{3}(t)=-\mu_{1}(t)+\mu_{3}(t)=-\sqrt{2} \phi_{1}(t)+\sqrt{1} \phi_{3}(t)
\end{aligned}
$$

6.40. Once we have constructed ${ }^{20}$ the set of, say $N$, orthonormal waveforms $\left\{\phi_{i}(t), i=1,2, \ldots, N\right\}$, we can express the signals $s_{i}(t)$ as linear combinations of the $N$ orthonormal basis functions $\phi_{i}(t)$. Thus, we may write

$$
\begin{equation*}
s_{j}(t)=\sum_{i=1}^{N} s_{i}^{(j)} \phi_{i}(t) \tag{35}
\end{equation*}
$$

where the constants (weights)

$$
\begin{equation*}
s_{i}^{(j)}=\left\langle s_{j}(t), \phi_{i}(t)\right\rangle \tag{36}
\end{equation*}
$$

[^0]Note that $s_{i}^{(j)} \phi_{i}(t)=\left\langle s_{j}(t), \phi_{i}(t)\right\rangle \phi_{i}(t)$ can be geometrically interpreted as the projection of the signal $s_{j}(t)$ onto the $i$ th axis, $\phi_{i}(t)$.

Based on (35), each signal may be represented by the vector (or sequence)

$$
\begin{equation*}
\mathbf{s}^{(j)}=\left(s_{1}^{(j)}, s_{2}^{(j)}, \ldots, s_{N}^{(j)}\right)^{T} \tag{37}
\end{equation*}
$$

or, equivalently, as a point in the $N$-dimensional (in general, complex) signal space.

The (mathematical/conceptual) conversion/mapping from waveform to it corresponding vector in (37) and (36) is shown in Figure 28a. The inverse mapping from vector to waveform in (35) is shown in Figure 28b.

(a)

(b)

Figure 28: Waveform to vector (a), and vector to waveform (b) mappings.
Example 6.41. For the four waveforms in Example 6.39 and the orthonormal basis derived from GSOP,

$$
\begin{aligned}
& s_{1}(t)=\underline{\sqrt{2}} \phi_{1}(t)+\underline{0} \phi_{2}(t)+\underline{0} \phi_{3}(t) \Rightarrow \mathbf{s}^{(1)}=\left(\begin{array}{c}
\sqrt{2} \\
0 \\
0
\end{array}\right) \\
& s_{2}(t)=\underline{0} \phi_{1}(t)+\underline{\sqrt{2}} \phi_{2}(t)+\underline{0} \phi_{3}(t) \Rightarrow \mathbf{s}^{(2)}=\left(\begin{array}{c}
0 \\
\sqrt{2} \\
0
\end{array}\right) \\
& s_{3}(t)=\underline{\sqrt{2}} \phi_{1}(t)+\underline{0} \phi_{2}(t)+\underline{1} \phi_{3}(t) \Rightarrow \mathbf{s}^{(3)}=\left(\begin{array}{c}
\sqrt{2} \\
0 \\
1
\end{array}\right) \\
& s_{4}(t)=-\underline{\sqrt{2}} \phi_{1}(t)+\underline{0} \phi_{2}(t)+\underline{1} \phi_{3}(t) \Rightarrow \mathbf{s}^{(4)}=\left(\begin{array}{c}
-\sqrt{2} \\
0 \\
1
\end{array}\right)
\end{aligned}
$$

Definition 6.42. From 6.40, a set of $M$ signals $\left\{s_{j}(t), j=1,2, \ldots, M\right\}$ can be represented by a set of $M$ vectors $\left\{\mathbf{s}^{(j)}\right\}$ in the $N$-dimensional space. The corresponding set of vectors is called the signal space representation, or constellation, of $\left\{s_{j}(t), j=1,2, \ldots, M\right\}$.
6.43. From the orthonormality of the basis, we have
(a) the inner product of two signals is equal to the inner product of the corresponding vectors:

$$
\left\langle s_{i}(t), s_{j}(t)\right\rangle=\left\langle\mathbf{s}^{(i)}, \mathbf{s}^{(j)}\right\rangle .
$$

(b) $E_{j} \equiv E_{s^{(j)}}=\left\|s_{j}(t)\right\|^{2}=\left\|\mathbf{s}^{(j)}\right\|^{2}$.
6.44. It should be emphasized, however, that the functions $\left\{\phi_{i}(t)\right\}$ obtained from the Gram-Schmidt procedure are not unique. If we alter the order in which the orthogonalization of the signals $\left\{s_{j}(t)\right\}$ is performed, the orthonormal waveforms will be different and the corresponding vector representation of the signals $\left\{s_{j}(t)\right\}$ will depend on the resulting orthonormal functions $\left\{\phi_{i}(t)\right\}$. Nevertheless, the dimensionality of the signal space $(N)$ will not change, and the vectors $\mathbf{s}^{(j)}$ will retain their geometric configuration; i.e., their lengths and their inner products will be invariant to the choice of the orthonormal functions $\left\{\phi_{i}(t)\right\}$.


[^0]:    ${ }^{20}$ We have shown how this set can be constructed from GSOP. However, in practice, this set may be derived from different procedure.

